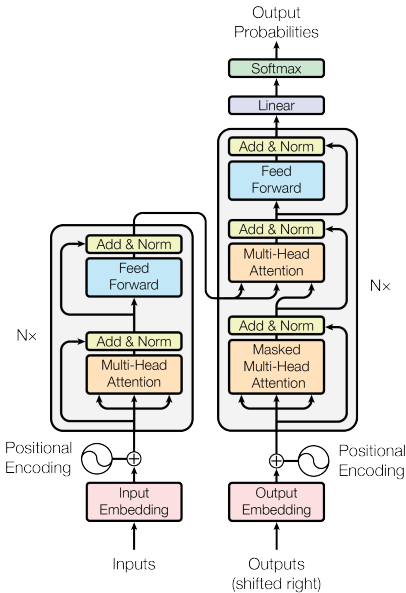


Analog Neural Networks and Translinear Circuits

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Attention Is All You Need



$$\mathbf{y} = \sigma \left(\begin{bmatrix} w_{11} & w_{12} & \dots & w_{1n} \\ w_{21} & w_{22} & \dots & w_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ w_{m1} & w_{m2} & \dots & w_{mn} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + \begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_m \end{bmatrix} \right)$$

$$OA_{(x,y,k)} = f \left(\sum_{i=0}^{R-1} \sum_{j=0}^{S-1} \sum_{c=0}^{C-1} IA_{(x+i,y+j,c)} \times W_{(i,j,c,k)} \right)$$

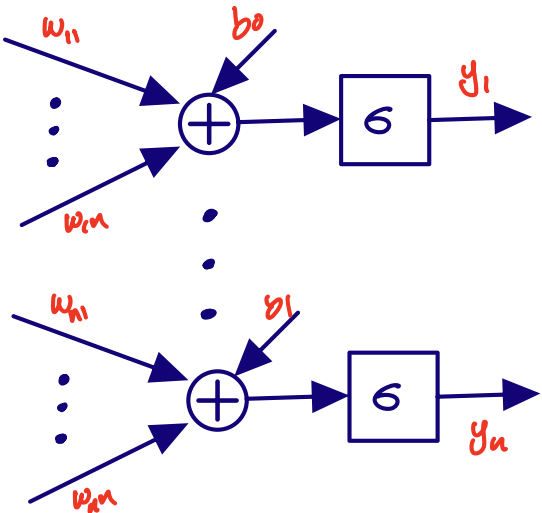
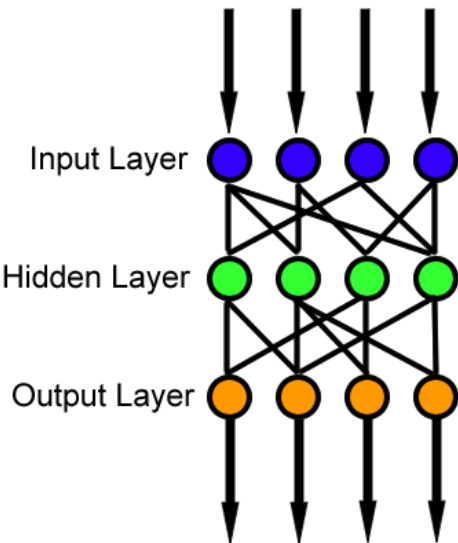
Assume N neurons

- N multiplications per neuron
- N + 1 additions per neuron
- 1 sigmoid per neuron

Neural Nets 3blue1brown

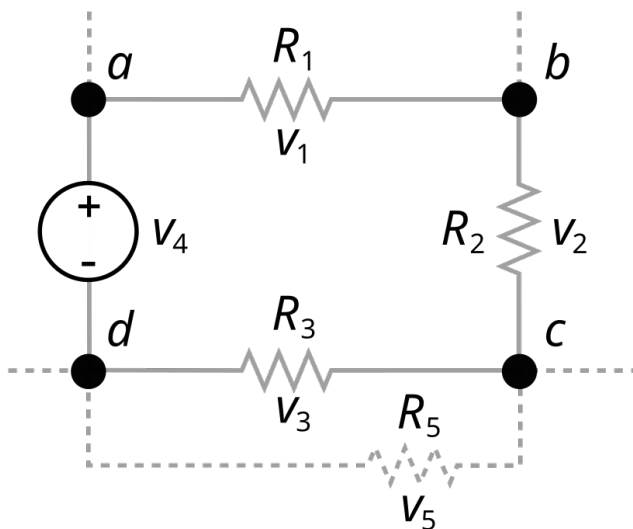
$$a_{l+1} = \sigma(W_l a_l + b_l)$$

A NN consists of addition, multiplication, and a non-linear function

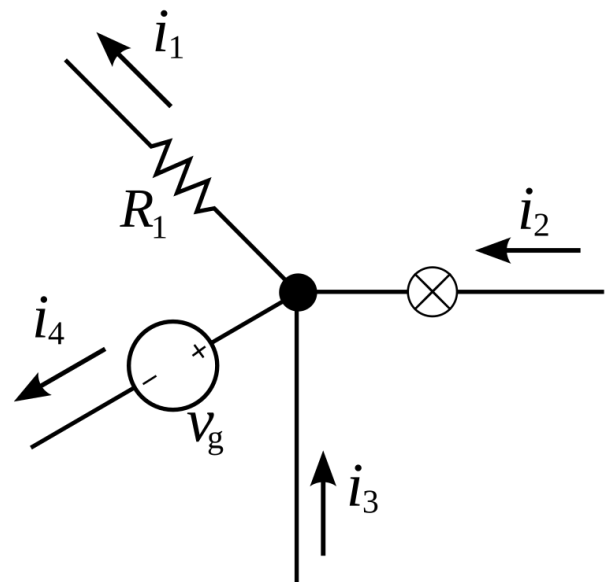


A. Kirchhoff's voltage law

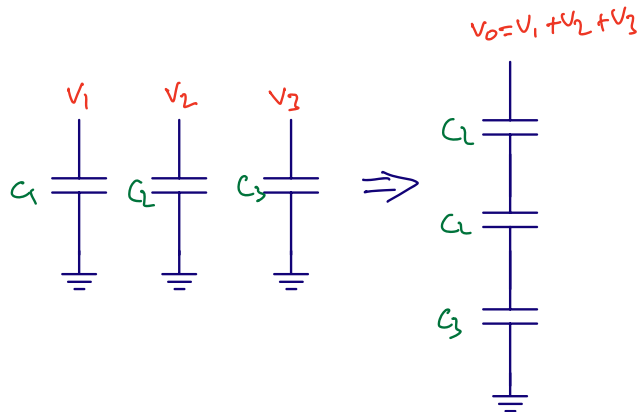
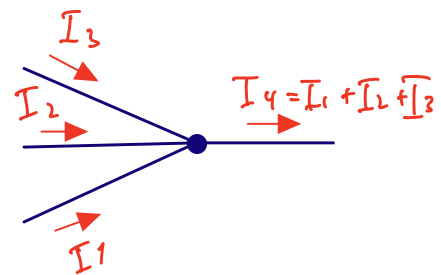
The directed sum of the potential differences around any closed loop is zero



$$V_1 + V_2 + V_3 + V_4 = 0$$



$$i_1 + i_2 + i_3 + i_4 = 0$$

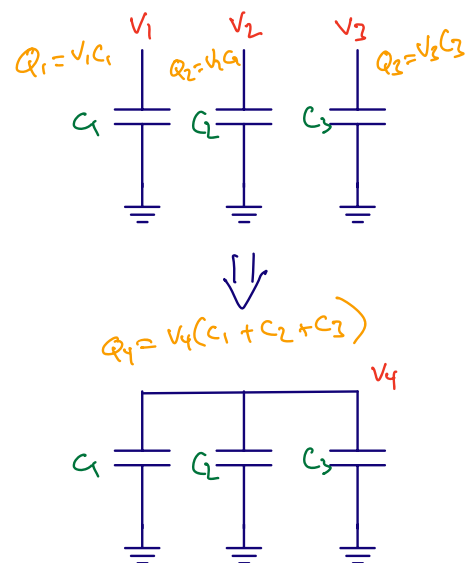


B. Kirchhoff's current law

The algebraic sum of currents in a network of conductors meeting at a point is zero

C. Charge conservation

See [Charge conservation](#) on Wikipedia



$$Q_4 = Q_1 + Q_2 + Q_3$$

$$V_4 = \frac{C_1 V_1 + C_2 V_2 + C_3 V_3}{C_1 + C_2 + C_3}$$

I. MULTIPLICATION

A. Digital capacitance

$$V_4 = \frac{C_1 V_1 + C_2 V_2 + C_3 V_3}{C_1 + C_2 + C_3}$$

$$V_O = \frac{C_1}{C_{TOT}} V_1 + \dots + \frac{C_N}{C_{TOT}} V_N$$

Make capacitors digitally controlled, then

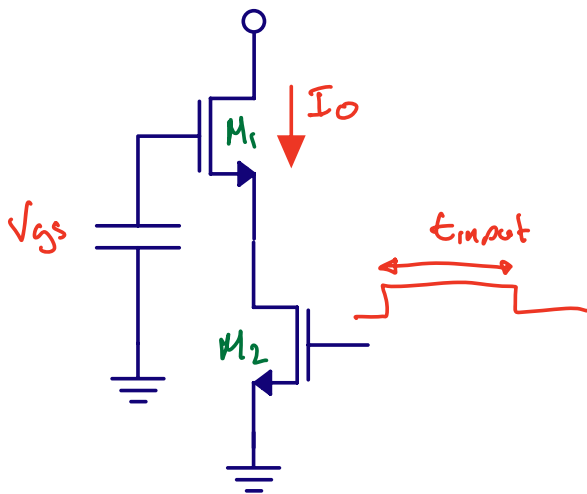
$$w_1 = \frac{C_1}{C_{TOT}}$$

Might have a slight problem with variable gain as a function of total capacitance

B. Mixing

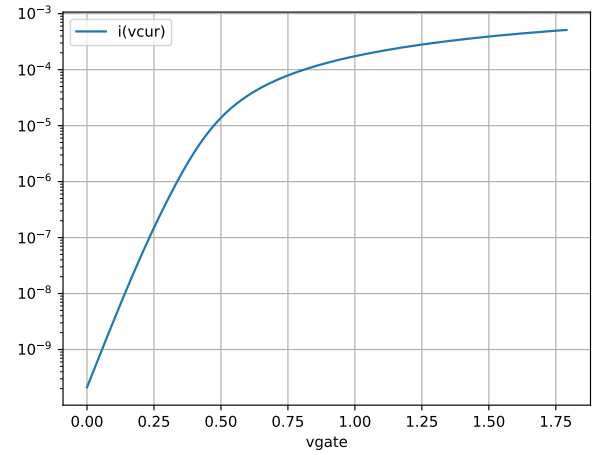
$$I_{M1} = G_m V_{GS}$$

$$I_o = I_{M1} t_{input}$$



C. Translinear principle

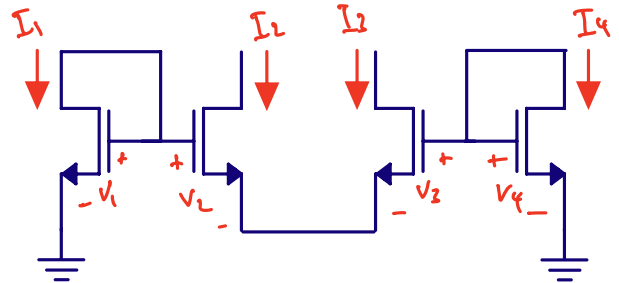
1) MOSFET in sub-threshold:



$$I = I_{D0} \frac{W}{L} e^{(V_{GS} - V_{th})/nU_T}, U_T = \frac{kT}{q}$$

$$I = \ell e^{V_{GS}/nU_T}, \ell = I_{D0} \frac{W}{L} e^{-V_{th}/nU_T}$$

$$V_{GS} = nU_T \ln \left(\frac{I}{\ell} \right)$$



$$V_1 + V_2 = V_3 + V_4$$

$$nU_T \left[\ln \left(\frac{I_1}{\ell_1} \right) + \ln \left(\frac{I_2}{\ell_2} \right) \right] = nU_T \left[\ln \left(\frac{I_3}{\ell_3} \right) + \ln \left(\frac{I_4}{\ell_4} \right) \right]$$

$$\ln \left(\frac{I_1 I_2}{\ell_1 \ell_2} \right) = \ln \left(\frac{I_3 I_4}{\ell_3 \ell_4} \right)$$

$$\frac{I_1 I_2}{\ell_1 \ell_2} = \frac{I_3 I_4}{\ell_3 \ell_4}$$

$$I_1 I_2 = I_3 I_4, \text{ if } \ell_1 \ell_2 = \ell_3 \ell_4$$

$$I_1 I_2 = I_3 I_4$$

$$I_1 = I_a, I_2 = I_b + i_b, I_3 = I_b, I_4 = I_a + i_a$$

$$I_a(I_b + i_b) = I_b(I_a + i_a)$$

$$I_a I_b + I_a i_b = I_b I_a + I_b i_a$$

$$i_b = \frac{I_b}{I_a} i_a$$

$$\ell_1 \ell_2 = \ell_3 \ell_4$$

$$\ell_1 = I_{D0} \frac{W}{L} e^{-V_{th}/nU_T}$$

$$\ell_2 = I_{D0} \frac{W}{L} e^{-(V_{th} \pm \sigma_{th})/nU_T} = \ell_1 e^{\pm \sigma_{th}/nU_T}$$

$$\sigma_{th} = \frac{a_{vt}}{\sqrt{WL}}$$

$$\frac{\ell_2}{\ell_1} = e^{\pm \frac{a_{vt}}{\sqrt{WL}}/nU_T}$$

2) Demo: JNW_SV_SKY130A

II. WANT TO LEARN MORE?

An Always-On 3.8 uJ/86 % CIFAR-10 Mixed-Signal Binary CNN Processor With All Memory on Chip in 28-nm CMOS

CAP-RAM: A Charge-Domain In-Memory Computing 6T-SRAM for Accurate and Precision-Programmable CNN Inference

ARCHON: A 332.7TOPS/W 5b Variation-Tolerant Analog CNN Processor Featuring Analog Neuronal Computation Unit and Analog Memory

IMPACT: A 1-to-4b 813-TOPS/W 22-nm FD-SOI Compute-in-Memory CNN Accelerator Featuring a 4.2-POPS/W 146-TOPS/mm² CIM-SRAM With Multi-Bit Analog Batch-Normalization



Carsten Wulff received the M.Sc. and Ph.D. degrees in electrical engineering from the Department of Electronics and Telecommunication, Norwegian University of Science and Technology (NTNU), in 2002 and 2008, respectively. During his Ph.D. work at NTNU, he worked on open-loop sigma-

delta modulators and analog-to-digital converters in nanoscale

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