

SPICE

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I. SPICE

Status: 0.3

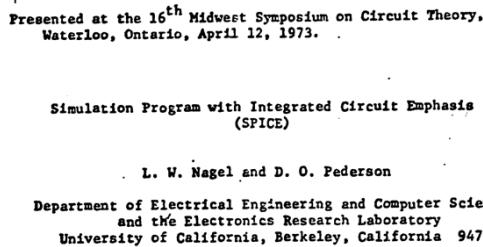
II. SIMULATION PROGRAM WITH INTEGRATED CIRCUIT EMPHASIS

To manufacture an integrated circuit we have to be able to predict how it's going to work. The only way to predict is to rely on our knowledge of physics, and build models of the real world in our computers.

One simulation strategy for a model of the real world, which absolutely every single integrated circuit in the world has used to come into existence, is SPICE.

Published in 1973 by Nagel and Pederson

[SPICE \(Simulation Program with Integrated Circuit Emphasis\)](#)



Abstract
A new circuit simulation program, SPICE, is described. The simulation capabilities of nonlinear dc analysis, small signal analysis, and nonlinear transient analysis are combined in a nodal analysis program to yield a reasonably general purpose electronic circuit simulation program. Particular emphasis is placed upon the circuit models for the BJT and the FET which are implemented in SPICE.

A. Today

There are multiple SPICE programs that has been written, but they all work in a similar fashion. There are expensive ones, closed source, and open source.

Some are better at dealing with complex circuits, some are faster, and some are more accurate. If you don't have money, then start with ngspice.

[Commercial Cadence Spectre Siemens Eldo Synopsys HSPICE](#)

[Free Aimspice Analog Devices LTspice](#)

[Open Source ngspice](#)

B. But

All SPICE simulators understand the same language (yes, even spectre can speak SPICE). We write our testbenches in a text

file, and give it to the SPICE program. That's the same for all programs. Some may have built fancy GUI's to hide the fact that we're really writing text files, but text files is what is under the hood.

Pretty much the same usage model as 48 years ago

```
<spice program> testbench.cir
```

for example

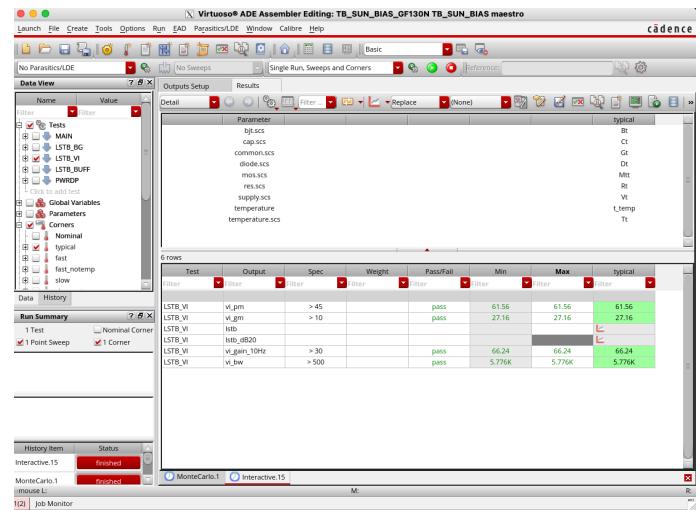
```
ngspice testbench.cir
```

Or in the most expensive analog tool (Cadence Spectre)

```
spectre input.scs +escchars +log ..//psf/spectre.out
-format psfxml -raw ..//psf +aps +lqtimeout 900 -maxw 5
-maxn 5 -env ade -ahdllibdir
/tmp/wulff/virtuoso/TB_SUN_BIAS_GF130N/TB_SUN_BIAS/maestro/
results/maestro/Interactive.15/sharedData/CDS/ahdl/input.ahdlSimDB
+logstatus
```

The expensive tools have built graphical user interface around the SPICE simulator to make it easier to run multiple scenarios.

Corner	Typical	Fast	Slow	All
Mosfet	Mtt	Mff	Mss	Mff,Mfs,Msf,Mss
Resistor	Rt	Rl	Rh	Rl,Rh
Capacitors	Ct	Cl	Ch	Cl,Ch
Diode	Dt	Df	Ds	Df,Ds
Bipolar	Bt	Bf	Bs	Bf,Bs
Temperature	Tt	Th,Tl	Th,Tl	Th,Tl
Voltage	Vt	Vh,Vl	Vh,Vl	Vh,Vl



I'm a fan of launching multiple simulations from the command line. I don't like GUI's. As such, I wrote [cicsim](#), and that's what I use in the video and demo.

C. Sources

The SPICE language is a set of conventions for how to write the text files. In general, it's one line, one command (although, lines can be continued with a +).

I'm not going to go through an extensive tutorial in this document, and there are dialects with different SPICE programs. You'll find more info at [ngspice](#)

1) Independent current sources: Infinite output impedance, changing voltage does not change current

```
I<name> <from> <to> dc <number> ac <number>
```

```
I1 0 VDN dc In
I2 VDP 0 dc Ip
```

2) Independent voltage source: Zero output impedance, changing current does not change voltage

```
V<name> <+> <-> dc <number> ac <number>
```

```
V2 VSS 0 dc 0
V1 VDD 0 dc 1.5
```

D. Passives

Resistors

```
R<name> <node 1> <node 2> <value>
```

```
R1 N1 N2 10k
R2 N2 N3 1Meg
R3 N3 N4 1G
R4 N4 N5 1T
```

Capacitors

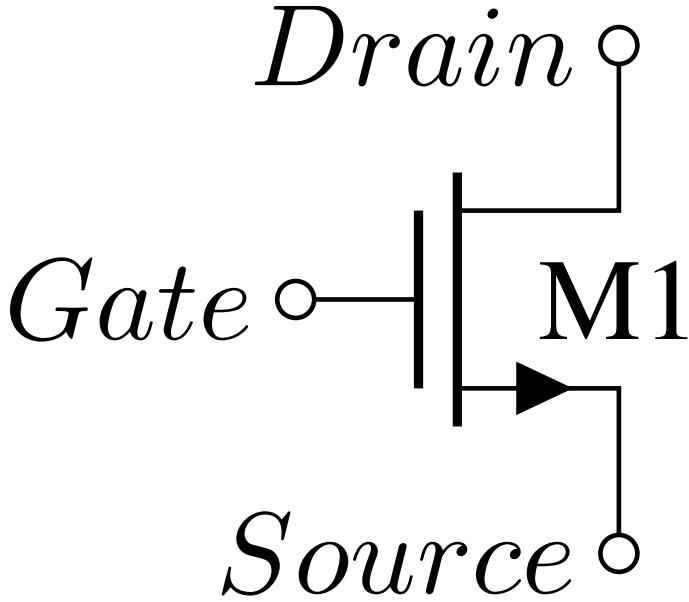
```
C<name> <node 1> <node 2> <value>
```

```
C1 N1 N2 1a
C2 N1 N2 1f
C4 N1 N2 1p
C3 N1 N2 1n
C5 N1 N2 1u
```

E. Transistor Models

Needs a model file the transistor model

BSIM (Berkeley Short-channel IGFET Model) <http://bsim.berkeley.edu/models/bsim4/>



284 parameters in **BSIM 4.5**

```
.MODEL N1 NMOS LEVEL=14 VERSION=4.5.0 BINUNIT=1
PARAMCHK=1 MOBMOD=0 CAPMOD=2 IGCMD=1 IGBMOD=1
```

```
GEOMOD=1 DIOMOD=1 RDSMOD=0 RBODYMOD=0 RGATEMOD=3
PERMOD=1 ACNQSMOD=0 TRNQSMOD=0 TEMPMOD=0 TNOM=27
TOXE=1.8E-009 TOXP=10E-010 TOXM=1.8E-009 DTOX=8E-10
EPSROX=3.9 WINT=5E-009 LINT=1E-009 LL=0 WL=0 LLN=1
WLN=1 LW=0 WW=0 LWN=1 WWN=1 LWL=0 WWL=0 XPART=0
TOXREF=1.4E-009 SAREF=5E-6 SBREF=5E-6 WL0D=2E-6
KUO=-4E-6 KVSA=0.2 KVTH0=-2E-8 TKUO=0.0 LLODKUO=1.1
WL0DKUO=1.1 LL0DVTH=1.0 WL0DVTH=1.0 LKUO=1E-6
WKUO=1E-6 PKUO=0.0 LKVTH0=1.1E-6 WKVTH0=1.1E-6
PKVTH0=0.0 STK2=0.0 LODK2=1.0 STETA0=0.0 LODETA0=1.0
LAMBDA=4E-10 VSAT=1.1E 005 VTL=2.0E5 XN=6.0 LC=5E-9
RN0IA=0.577 RNOIB=0.37 LINTNOI=1E-009 WPEMOD=0
WEB=0.0 WEC=0.0 KVTH0WE=1.0 K2WE=1.0 KU0WE=1.0 SCREF=5.0E-6
TVOFF=0.0 TVFBSDOFF=0.0 VT0H=0.25 K1=0.35 K2=0.05
K3=0 K3B=0 W0=2.5E-006 DVT0=1.8 DVT1=0.52 DVT2=-0.032
DVTOW=0 DVT1W=0 DVT2W=0 DSUB=2 MINV=0.05 VOFFL=0
DVTPO=1E-007 DVTPI=0.05 LP0=5.75E-008 LP0E=2.3E-010
XJ=2E-008 NGATE=5E 020 NDEP=2.8E 018 NSD=1E 020 PHIN=0
CDSC=0.0002 CDSCB=0 CDSCD=0 CIT=0 VOFF=-0.15 NFACT0=1.2
ETA0=0.05 ETAB=0 UC=-3E-011 VFB=-0.55 U0=0.032
UA=5.0E-011 UB=3.5E-018 A0=2 AGS=1E-020 A1=0 A2=1
B0=-1E-020 B1=0 KETA=0.04 DWG=0 DWB=0 PCLM=0.08
PDIBLC1=0.028 PDIBLC2=0.022 PDIBLCB=-0.005 DROUT=0.45
PVAG=1E-020 DELTA=0.01 PSCBE1=8.14E 008 PSCBE2=5E-008
RSH=0 RDWS=0 RSW=0 RDW=0 FPROUT=0.2 PDITS=0.2 PDITSD=0.23
PDITSL=2.3E 006 RSH=0 RDWS=50 RSW=150
RDW=150 RDWSMIN=0 RDWMIN=0 RSWMIN=0 PRWG=0 PRWB=6.8E-011
WR=1 ALPHAO=0.074 ALPHAI=0.005 BETA0=30 AGIDL=0.0002
BGIDL=2.1E 009 CGIDL=0.0002 EGIDL=0.8 AIGBACC=0.012
BIGBACC=0.0028 CIGBACC=0.002 NIGBACC=1 AIGBIN=0.014
BIGBIN=0.004 CIGBIN=0.004 EIGBIN=1.1 NIGBIN=3 AIGC=0.012
BIGC=0.0028 CIGC=0.002 AIGSD=0.012 BIGSD=0.0028 CIGSD=0.002 NIGC=1
POXEDGE=1 PIGCD=1 NTOX=1 VFBSDOFF=0.0 XRCRG1=12 XRCRG2=5
CGSO=6.238E-010 CGDO=6.238E-010 CGBO=2.56E-011 CGDL=2.495E-10
CGSL=2.495E-10 CKAPPAS=0.03 CKAPPAD=0.03 ACDE=1 MOIN=15
NOFF=0.9 VOFFCV=0.02 KT1=-0.37 KT1L=0.0 KT2=-0.042 UTE=-1.5
UAI=1E-009 UB1=-3.5E-019 UCI=0 PRT=0 AT=53000 FN0IMOD=1
TNOIMOD=0 JSS=0.0001 JSWS=1E-011 JSWGS=1E-010 NJS=1
IJTHSFWD=0.01 IJTHSREV=0.001 BVS=10 XJBVS=1 JSD=0.0001
JSWD=1E-011 JSWGD=1E-010 NJD=1 IJTHDFWD=0.01 IJTHDREV=0.001
BVD=10 XJBVD=1 PBS=1 CJS=0.0005 MJS=0.5 PBSWS=1 CJSWS=5E-010
MJSWS=0.33 PBSWGS=1 CJSWGS=3E-010 MJSWGS=0.33 PBD=1 CJD=0.0005
MJD=0.5 PBSWD=1 CJSWD=5E-010 MJSWD=0.33 PBSWGD=1
CJSWGD=5E-010 MJSWGD=0.33 TPB=0.005 TCJ=0.001 TPBSW=0.005
TCJSW=0.001 TPBSWG=0.005 TCJSWG=0.001 XTIS=3 XTID=3 DMCG=0E-006
DMCI=0E-006 DMDG=0E-006 DMCGT=0E-007 DWJ=0.0E-008 XGW=0E-007
XGL=0E-008 RSHG=0.4 GBMIN=1E-010 RBPB=5 RBPD=15 RBPS=15 RBDB=15
RBSB=15 NGCON=1 JTSS=1E-4 JTSD=1E-4 JTSSWS=1E-10 JTSSWD=1E-10
JTSSWGS=1E-7 JTSSWGD=1E-7 NJTS=10 NJTSS=20 NJTSSWG=6
VTSS=10 VTDS=10 VTSSWS=10 VTSSWD=10 VTSSWGS=2 VTSSWD=2
XTSS=0.02 XTSD=0.02 XTSSWS=0.02 XTSSWD=0.02 XTSSWG=0.02
XTSSWGD=0.02
```

F. Transistors

```
M<name> <drain> <gate> <source> <bulk> <modelname> [parameters]
```

```
M1 VDN VDN VSS VSS nmos W=0.6u L=0.15u
M2 VDP VDP VDD VDD pmos W=0.6u L=0.15u
```

G. Foundries

Each foundry has their own SPICE models because the transistor parameters depend on the exact physics of the technology!

<https://skywater-pdk.readthedocs.io/en/main/>

III. FIND RIGHT TRANSISTOR SIZES

Assume active (

$$V_{ds} > V_{eff}$$

in strong inversion, or

$$V_{ds} > 3V_T$$

in weak inversion). For diode connected transistors, that is always true.

Weak inversion:

$$I_D = I_{D0} \frac{W}{L} e^{V_{eff}/nV_T}$$

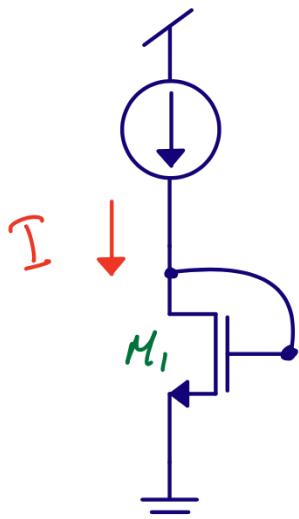
$$V_{eff} \propto \ln I_D$$

Strong inversion:

$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} V_{eff}^2$$

$$V_{eff} \propto \sqrt{I_D}$$

Operating region for a diode connected transistor only depends on the current



A. Use unit size transistors for analog design

$$W/L \approx [4, 6, 10]$$

, but should have space for two contacts

Use parallel transistors for larger W/L

Amplifiers

$$\Rightarrow L \approx 1.2 \times L_{min}$$

Current mirrors

$$\Rightarrow L \approx 4 \times L_{min}$$

Choose sizes that have been used by foundry for measurement to match SPICE model

B. What about gm/Id ?

Weak

$$\frac{g_m}{I_d} = \frac{1}{nV_T}$$

Strong

$$\frac{g_m}{I_d} = \frac{2}{V_{eff}}$$

C. Characterize the transistors

http://analogicus.com/cnr_atr_sky130nm/mos/CNRATR_NCH_2C1F2.html

IV. MORE INFORMATION

[Ngspice Manual](#)

[Installing tools](#)

V. ANALOG DESIGN

1. Define the problem, what are you trying to solve?
2. Find a circuit that can solve the problem (papers, books)
3. Find right transistor sizes. What transistors should be weak inversion, strong inversion, or don't care?
4. Check operating region of transistors (.op)
5. Check key parameters (.dc, .ac, .tran)
6. Check function. Exercise all inputs. Check all control signals
7. Check key parameters in all corners. Check mismatch (Monte-Carlo simulation)
8. Do layout, and check it's error free. Run design rule checks (DRC). Check layout versus schematic (LVS)
9. Extract parasitics from layout. Resistance, capacitance, and inductance if necessary.
10. On extracted parasitic netlist, check key parameters in all corners and mismatch (if possible).
11. If everything works, then your done.

On failure, go back

VI. DEMO

https://github.com/analogicus/jnw_spice_sky130A/tree/main



Carsten Wulff received the M.Sc. and Ph.D. degrees in electrical engineering from the Department of Electronics and Telecommunication, Norwegian University of Science and Technology (NTNU), in 2002 and 2008, respectively. During his Ph.D. work at NTNU, he worked on open-loop sigma-delta modulators and analog-to-digital converters in nanoscale CMOS technologies. In 2006-2007, he was a Visiting Researcher with the Department of Electrical and Computer Engineering, University of Toronto, Toronto, ON, Canada. Since 2008 he's been with Nordic Semiconductor in various roles, from analog designer, to Wireless Group Manager,

to currently Principle IC Scientist. He's also an Adjunct Associate Professor at NTNU. His present research interests includes analog and mixed-signal CMOS design, design of high-efficiency analog-to-digital converters and low-power wireless transceivers. He is the developer of Custom IC Compiler, a general purpose integrated circuit compiler.